

Assignment-3

$$f_c = 80 \text{ Hz}$$

$$f_s = 200 \text{ Hz}$$

$$N = 2$$

Normalized ^{cut off} angular frequency, $\omega_c = \frac{f_c}{f_s} 2\pi$

$$= \frac{80}{200} \times 2\pi$$

$$= 0.8\pi \text{ radians/s}$$

$$\text{So, } T = 1$$

The prewarped s-domain cut off frequency,

$$\Omega_c = \frac{2}{T} \tan\left(\frac{\omega_c}{2}\right)$$

$$= 2 \tan(0.4\pi)$$

$$= 2 \times 3.0777$$

$$= 6.1554 \text{ radians/s}$$

We know,

$$S_k = \Omega_c \exp\left[j\pi\left(\frac{1}{2} + \frac{2k-1}{2N}\right)\right], \quad k = 1, 2, \dots, 2N$$

$$\text{So, } s_1 = 6.1554 \exp\left[j\pi\left(\frac{1}{2} + \frac{1}{4}\right)\right]$$

$$= 6.1554 \exp(j0.75\pi)$$

$$= 6.1554 [\cos(0.75\pi) + j \sin(0.75\pi)]$$

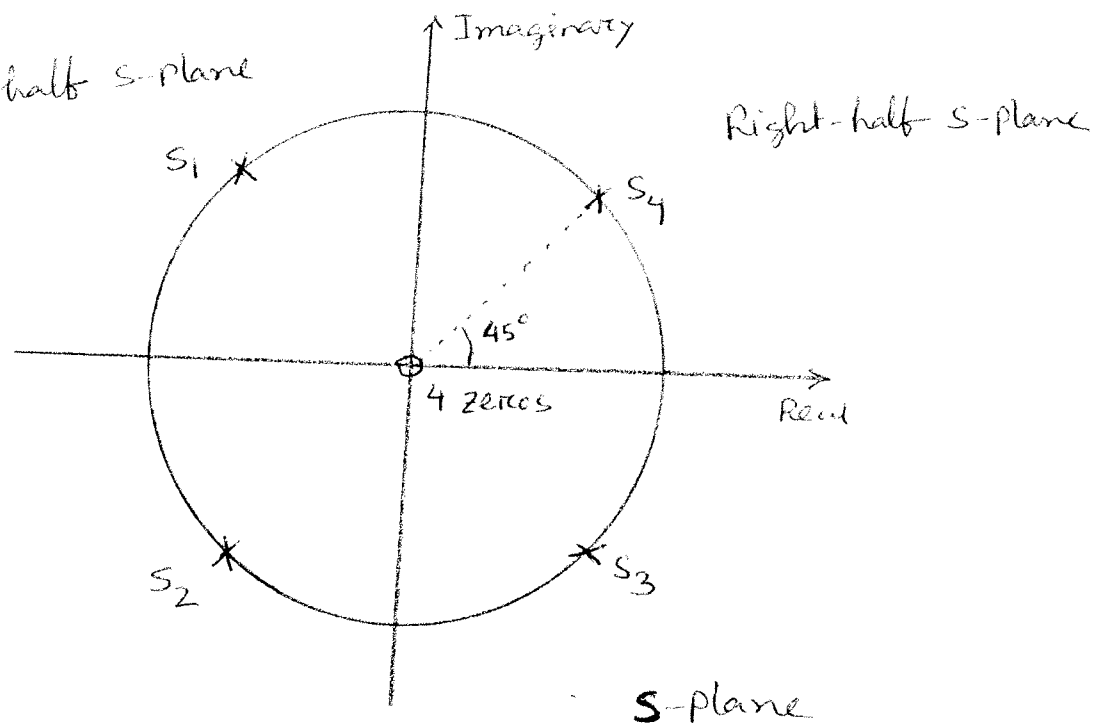
$$= -4.3525 + j4.3525$$

$$\begin{aligned}
 s_2 &= 6.1554 \exp\left[j\pi\left(\frac{1}{2} + \frac{3}{4}\right)\right] \\
 &= 6.1554 \exp(j1.25\pi) \\
 &= -4.3525 - j4.3525
 \end{aligned}$$

$$\begin{aligned}
 s_3 &= 6.1554 \exp\left[j\pi\left(\frac{1}{2} + \frac{5}{4}\right)\right] \\
 &= 6.1554 \exp(j1.75\pi) \\
 &= 4.3525 - j4.3525
 \end{aligned}$$

$$\begin{aligned}
 s_4 &= 6.1554 \exp(j2.25\pi) \\
 &= 4.3525 + j4.3525
 \end{aligned}$$

The pole-zero diagram will be,
left-half s-plane



Only s_1 and s_2 poles are on the ~~left hand side~~ left-half of the s -plane. So, the transfer function of the filter is,

$$H_a(s) = \frac{G}{(s + 4.3525 - j4.3525)(s + 4.3525 + j4.3525)}$$

where G is the gain factor.

$$H_a(s) = \frac{G}{s^2 + 8.705s + 37.8885}$$

Normalizing for unity gain at DC ($s=0$), the gain factor,

$$G = \frac{1}{H_a(s)|_{s=0}} = \frac{1}{\frac{1}{0+0+37.8885}} = 37.8885$$

So, the normalized filter in Laplace domain,

$$H_a(s) = \frac{37.8885}{s^2 + 8.705s + 37.8885}$$

The bilinear transformation

$$s = \frac{2}{T} \left[\frac{1 - z^{-1}}{1 + z^{-1}} \right] = 2 \left[\frac{1 - z^{-1}}{1 + z^{-1}} \right]$$

Applying bilinear transformation,

$$\begin{aligned}
 H(z) &= \frac{37.8885}{4 \frac{(1-z^{-1})^2}{(1+z^{-1})^2} + 8.705 \times 2 \frac{(1-z^{-1})}{(1+z^{-1})} + 37.8885} \\
 &= \frac{37.8885(1+z^{-1})^2}{4(1-z^{-1})^2 + 17.41(1-z^{-1})(1+z^{-1}) + 37.8885(1+z^{-1})^2} \\
 &= \frac{37.8885(1+2z^{-1}+z^{-2})}{4(1-2z^{-1}+z^{-2}) + 17.41(1-z^{-2}) + 37.8885(1+2z^{-1}+z^{-2})} \\
 &= \frac{37.8885(1+2z^{-1}+z^{-2})}{4 - 8z^{-1} + 4z^{-2} + 17.41 - 17.41z^{-2} + 37.8885 + 75.777z^{-1} + 37.8885z^{-2}} \\
 &= \frac{37.8885(1+2z^{-1}+z^{-2})}{59.2985 + 67.777z^{-1} + 24.4785z^{-2}} \\
 &= \frac{0.6389(1+2z^{-1}+z^{-2})}{1 + 1.1430z^{-1} + 0.4128z^{-2}}
 \end{aligned}$$

For unity gain at DC ($z=1$),

$$\begin{aligned}
 H(z)|_{z=1} &= \frac{0.6389(1+2+1)}{1+1.1430+0.4128} \\
 &= \frac{2.5556}{2.5558}
 \end{aligned}$$

$$\approx 1$$

So, the filter $H(z)$ is already normalized for unity gain at dc.

The Filter coefficients are,

$$[a] = [1 \quad 1.1430 \quad 0.4128]$$

$$[b] = [0.6389 \quad 1.2778 \quad 0.6389]$$

The pole-zero plot

$$H(z) = \frac{0.6389(1+z^{-1})^2}{1 + 1.1430z^{-1} + 0.4128z^{-2}}$$

The zeros are,

$$0.6389(1+z^{-1})^2 = 0$$

$$\Rightarrow (z+1)^2 = 0$$

$$z_1 = -1, z_2 = -1$$

~~The poles are,~~

~~$$p_1 = -1.1430 + \sqrt{(1.1430)^2 - 4 \times 1 \times 0.4128}$$~~

For the poles,

$$1 + 1.1430z^{-1} + 0.4128z^{-2} = 0$$

$$\Rightarrow z^2 + 1.1430z + 0.4128 = 0$$

$$p_1 = \frac{-1.1430 + \sqrt{(1.1430)^2 - 4 \times 1 \times 0.4128}}{2 \times 1}$$

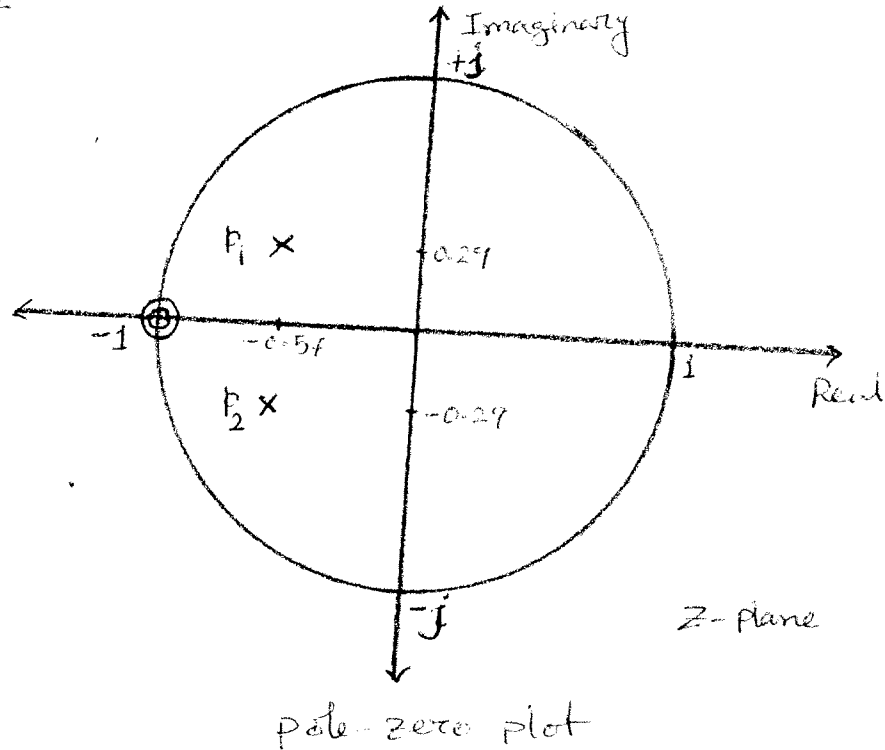
$$= \frac{-1.1430 + \sqrt{-0.3448}}{2}$$

$$= \frac{-1.1430 + 0.5872j}{2}$$

$$= -0.5715 + j0.2936$$

Similarly,

$$p_2 = -0.5715 - j0.2936$$



MATLAB

Butter-worth filter coefficients

$$[b] = [0.6389 \quad 1.2779 \quad 0.6389];$$

$$[a] = [1 \quad 1.1430 \quad 0.4128];$$

They are the same as calculated.

The frequency response of the filter is attached.

Fig: Frequency response of the filter

